

Communication Electronics

Lecture 4:

Antennas and propagation
of electromagnetic waves

Coordinate systems

Cartesian

$$3D = (x, y, z)$$

$$-\infty < x \text{ [m]} < +\infty$$

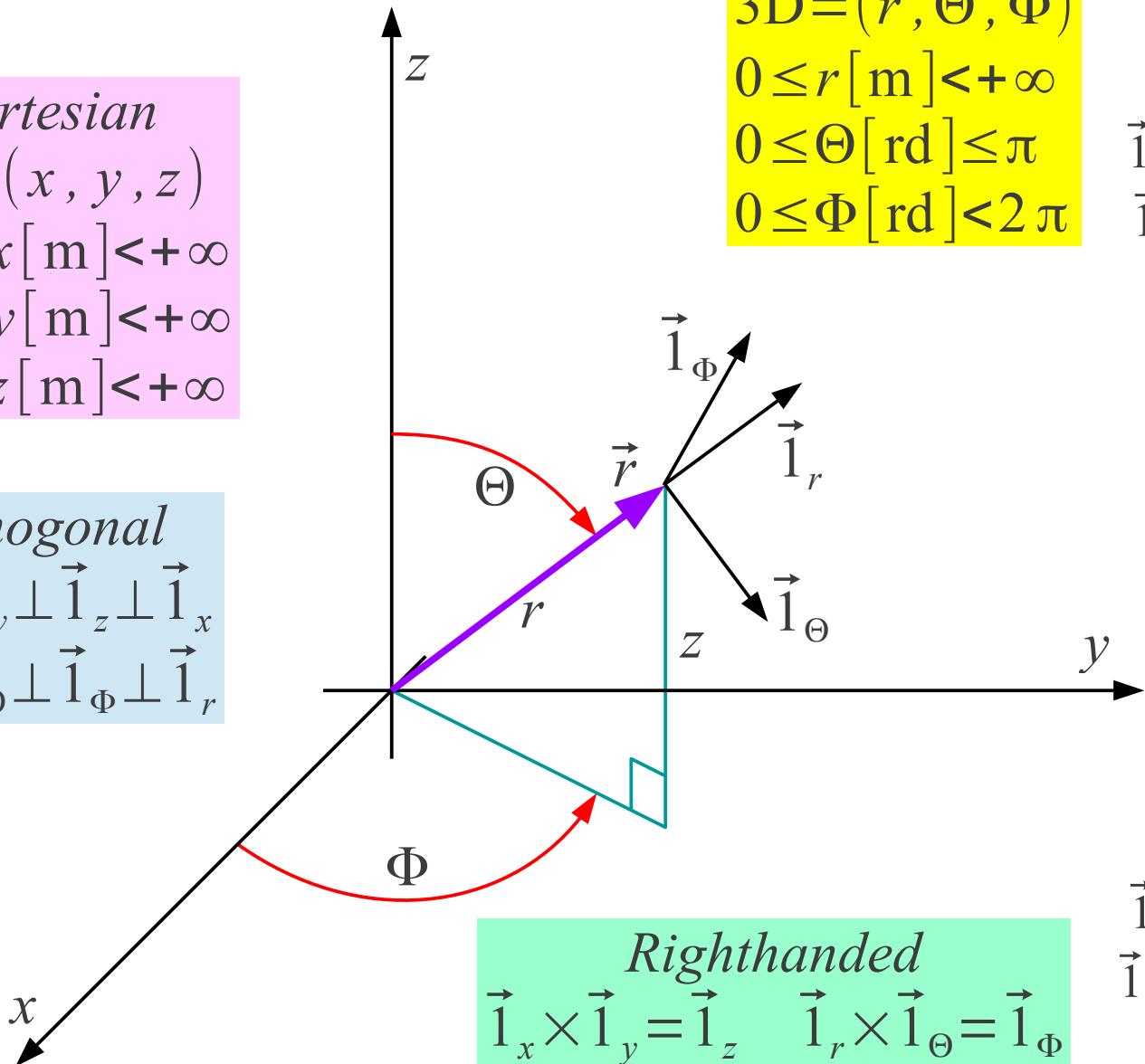
$$-\infty < y \text{ [m]} < +\infty$$

$$-\infty < z \text{ [m]} < +\infty$$

Orthogonal

$$\vec{1}_x \perp \vec{1}_y \perp \vec{1}_z \perp \vec{1}_x$$

$$\vec{1}_r \perp \vec{1}_\Theta \perp \vec{1}_\Phi \perp \vec{1}_r$$



Righthanded

$$\vec{1}_x \times \vec{1}_y = \vec{1}_z \quad \vec{1}_r \times \vec{1}_\Theta = \vec{1}_\Phi$$

Spherical

$3D = (r, \Theta, \Phi)$

$0 \leq r \text{ [m]} < +\infty$

$0 \leq \Theta \text{ [rd]} \leq \pi$

$0 \leq \Phi \text{ [rd]} < 2\pi$

Conversion $(r, \Theta, \Phi) \rightarrow (x, y, z)$

$$x = r \sin \Theta \cos \Phi$$

$$y = r \sin \Theta \sin \Phi$$

$$z = r \cos \Theta$$

$$\vec{1}_x = \vec{1}_r \sin \Theta \cos \Phi + \vec{1}_\Theta \cos \Theta \cos \Phi - \vec{1}_\Phi \sin \Phi$$

$$\vec{1}_y = \vec{1}_r \sin \Theta \sin \Phi + \vec{1}_\Theta \cos \Theta \sin \Phi + \vec{1}_\Phi \cos \Phi$$

$$\vec{1}_z = \vec{1}_r \cos \Theta - \vec{1}_\Theta \sin \Theta$$

$$0 \leq \Theta \leq \pi \rightarrow \sin \Theta \geq 0$$

Conversion $(x, y, z) \rightarrow (r, \Theta, \Phi)$

$$r = \sqrt{x^2 + y^2 + z^2}$$

$$\Theta = \arccos(z / \sqrt{x^2 + y^2 + z^2})$$

$$\Phi = \arctan(y/x) \quad (\text{quadrant?})$$

$$\vec{1}_r = \vec{1}_x \sin \Theta \cos \Phi + \vec{1}_y \sin \Theta \sin \Phi + \vec{1}_z \cos \Theta$$

$$\vec{1}_\Theta = \vec{1}_x \cos \Theta \cos \Phi + \vec{1}_y \cos \Theta \sin \Phi - \vec{1}_z \sin \Theta$$

$$\vec{1}_\Phi = -\vec{1}_x \sin \Phi + \vec{1}_y \cos \Phi$$

Maxwell equations

$$\begin{array}{ll} \text{Ampère} & \nabla \times \vec{H} = \vec{J} + j\omega \vec{D} \\ \text{Faraday} & \nabla \times \vec{E} = -j\omega \vec{B} \\ \text{Gauss} & \nabla \cdot \vec{D} = \rho \end{array}$$

\vec{B} [Vs/m²] \equiv magnetic flux density
 \vec{D} [As/m²] \equiv electric displacement field
 \vec{E} [V/m] \equiv electric field intensity
 \vec{H} [A/m] \equiv magnetic field intensity
 \vec{J} [A/m²] \equiv conductive current density
 ρ [As/m³] \equiv electric charge density

$$\text{Time-harmonic derivative} \quad \frac{\partial}{\partial t} = j\omega$$

$$\text{Spatial derivatives} \quad \nabla = \vec{i}_x \frac{\partial}{\partial x} + \vec{i}_y \frac{\partial}{\partial y} + \vec{i}_z \frac{\partial}{\partial z}$$

Potentials

$$\begin{aligned} V[\text{V}] &\equiv \text{scalar potential} \\ \vec{A}[\text{Vs/m}] &\equiv \text{vector potential} \\ \vec{E} &= -j\omega \vec{A} - \nabla V \\ \vec{B} &= \nabla \times \vec{A} \end{aligned}$$

Wave equations (Lorenz gauge)

$$\begin{aligned} \Delta \vec{A} + k^2 \vec{A} &= -\mu \vec{J} \\ \Delta V + k^2 V &= -\frac{\rho}{\epsilon} \end{aligned}$$

Vector operations

$$\begin{aligned} \vec{A} &= \vec{i}_x A_x + \vec{i}_y A_y + \vec{i}_z A_z \\ \vec{A} \cdot \vec{B} &= A_x B_x + A_y B_y + A_z B_z \\ \vec{A} \times \vec{B} &= \begin{vmatrix} \vec{i}_x & \vec{i}_y & \vec{i}_z \\ A_x & A_y & A_z \\ B_x & B_y & B_z \end{vmatrix} \end{aligned}$$

$$\text{Laplace} \quad \Delta = \frac{\partial^2}{\partial x^2} + \frac{\partial^2}{\partial y^2} + \frac{\partial^2}{\partial z^2}$$

$$k = \omega \sqrt{\mu \epsilon} = \frac{2\pi}{\lambda} \equiv \text{wavenumber}$$

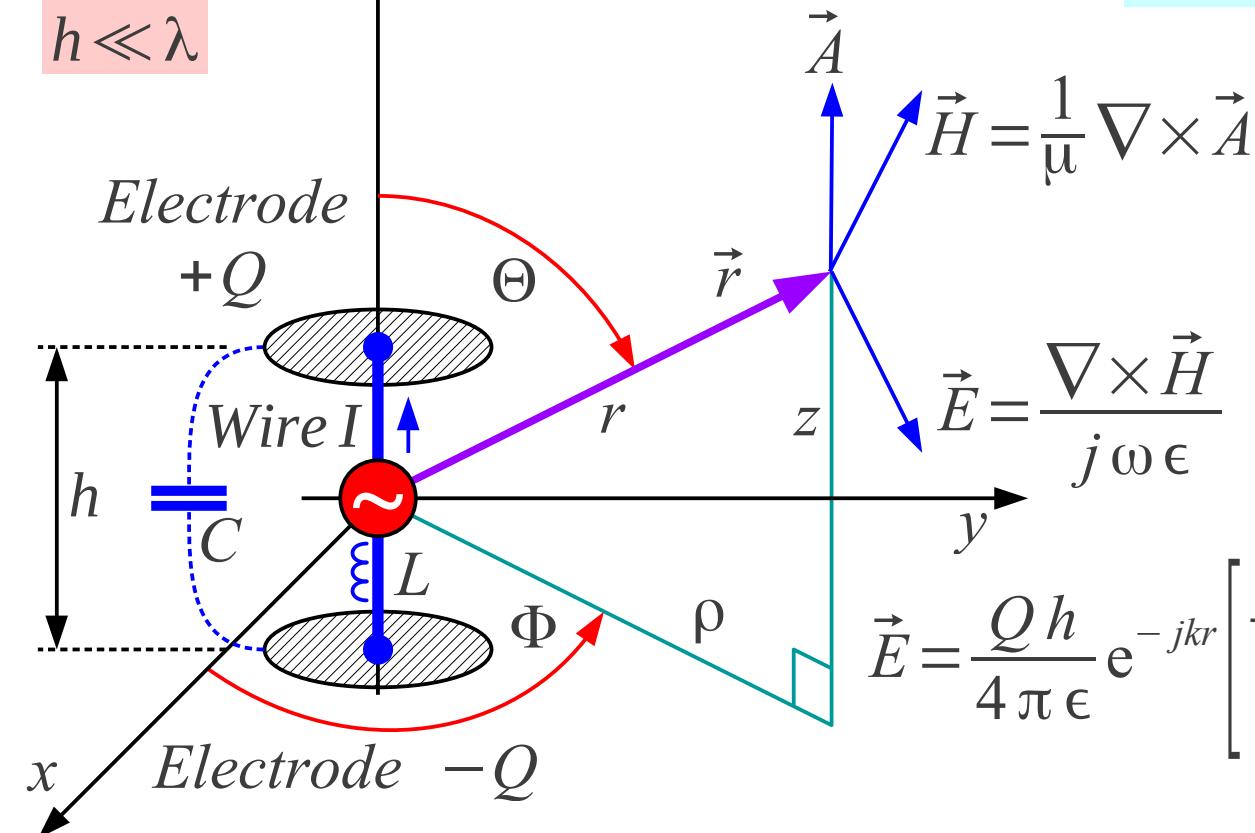
Retarded potentials

$$\begin{aligned} \text{Matter} \\ \vec{D} &= \epsilon \vec{E} \\ \vec{B} &= \mu \vec{H} \\ \vec{J} &= \gamma \vec{E} \end{aligned}$$

$$\vec{A}(\vec{r}) = \frac{\mu}{4\pi} \int_{V'} \vec{J}(\vec{r}') \frac{e^{-jk|\vec{r}-\vec{r}'|}}{|\vec{r}-\vec{r}'|} dV'$$

$$V(\vec{r}) = \frac{1}{4\pi\epsilon} \int_{V'} \rho(\vec{r}') \frac{e^{-jk|\vec{r}-\vec{r}'|}}{|\vec{r}-\vec{r}'|} dV'$$

$\omega \neq 0$
 $h \ll r$
 $h \ll \lambda$



$$\vec{A} \approx \vec{l}_z \frac{\mu I h}{4\pi} \frac{e^{-jkr}}{r} = (\vec{l}_r \cos \Theta - \vec{l}_\Theta \sin \Theta) \frac{\mu I h}{4\pi} \frac{e^{-jkr}}{r}$$

$$\vec{H} = \vec{l}_\Phi \frac{I h}{4\pi} e^{-jkr} \left(\frac{jk}{r} + \frac{1}{r^2} \right) \sin \Theta$$

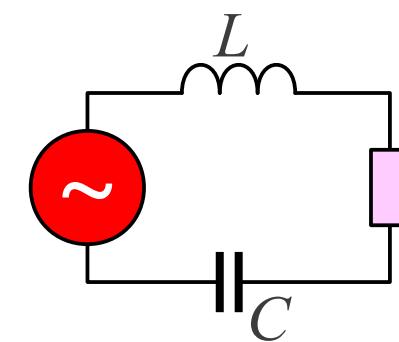
Continuity
 $I = j \omega Q$

$$\vec{E} = \frac{Q h}{4\pi \epsilon} e^{-jkr} \left[\vec{l}_r \left(\frac{jk}{r^2} + \frac{1}{r^3} \right) 2 \cos \Theta + \vec{l}_\Theta \left(-\frac{k^2}{r} + \frac{jk}{r^2} + \frac{1}{r^3} \right) \sin \Theta \right]$$

Radiation

$$\text{Re}[\vec{S}] = \text{Re} \left[\frac{1}{2} \vec{E} \times \vec{H}^* \right] = \vec{l}_r \frac{Z k^2}{32\pi^2} |I|^2 h^2 \frac{\sin^2 \Theta}{r^2}$$

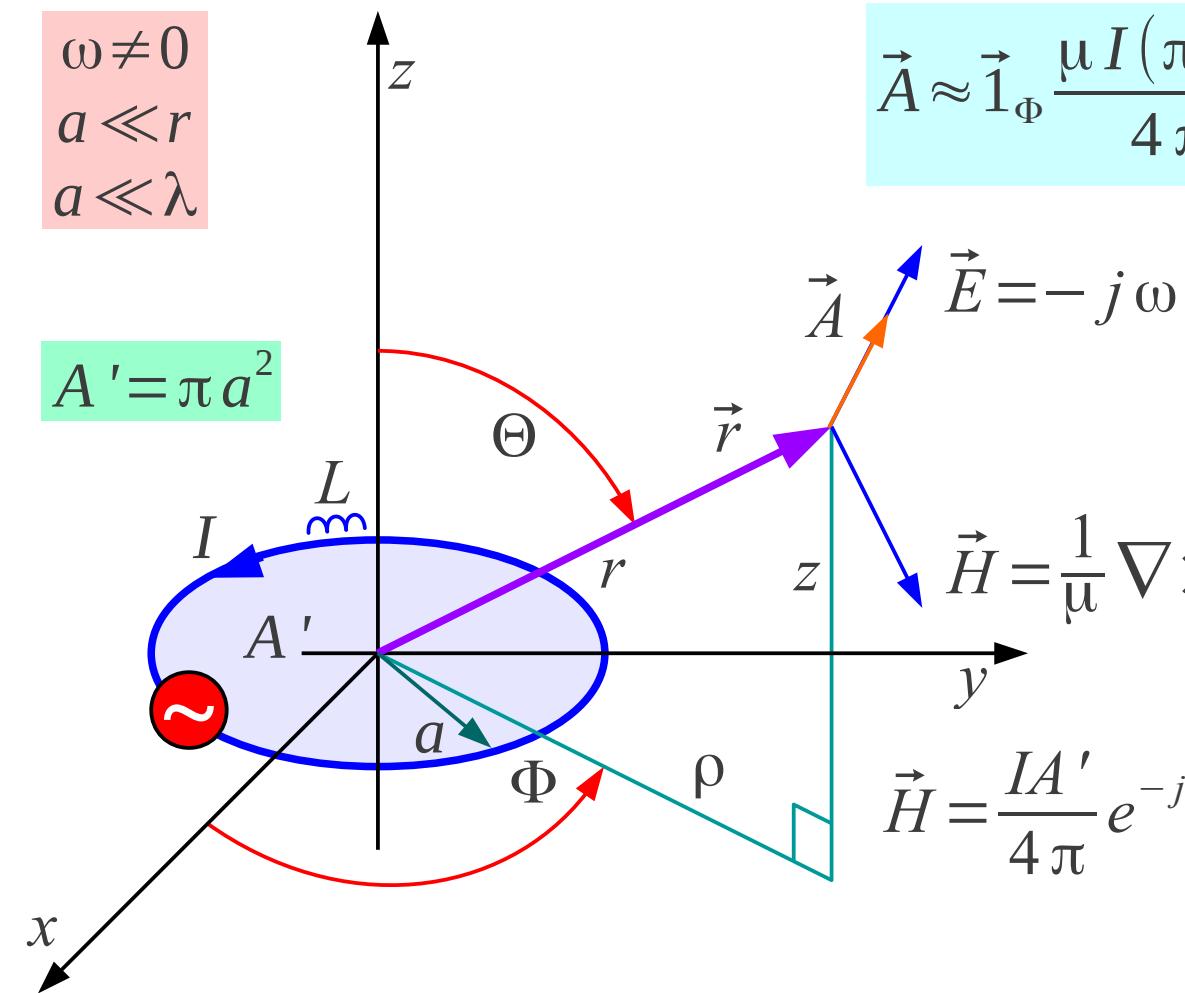
Small electric dipole



$$R = \frac{Z k^2 h^2}{6\pi} = \frac{2\pi Z}{3} \left(\frac{h}{\lambda} \right)^2$$

$$\begin{aligned}\omega &\neq 0 \\ a &\ll r \\ a &\ll \lambda\end{aligned}$$

$$A' = \pi a^2$$



$$\vec{A} \approx \vec{l}_\Phi \frac{\mu I (\pi a^2)}{4\pi} e^{-jkr} \left(\frac{jk}{r} + \frac{1}{r^2} \right) \sin \Theta$$

$$Q=0 \rightarrow V=0$$

$$\vec{E} = -j\omega \vec{A} - \nabla V = \vec{l}_\Phi \frac{ZIA'}{4\pi} e^{-jkr} \left(\frac{k^2}{r} - \frac{jk}{r^2} \right) \sin \Theta$$

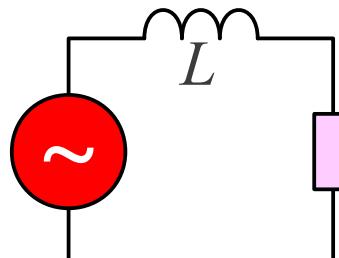
Radiation

$$\vec{H} = \frac{1}{\mu} \nabla \times \vec{A}$$

$$\vec{H} = \frac{IA'}{4\pi} e^{-jkr} \left[\vec{l}_r \left(\frac{jk}{r^2} + \frac{1}{r^3} \right) 2 \cos \Theta + \vec{l}_\Theta \left(-\frac{k^2}{r} + \frac{jk}{r^2} + \frac{1}{r^3} \right) \sin \Theta \right]$$

$$\text{Re}[\vec{S}] = \text{Re} \left[\frac{1}{2} \vec{E} \times \vec{H}^* \right] = \vec{l}_r \frac{Z k^4}{32 \pi^2} |I|^2 (A')^2 \frac{\sin^2 \Theta}{r^2}$$

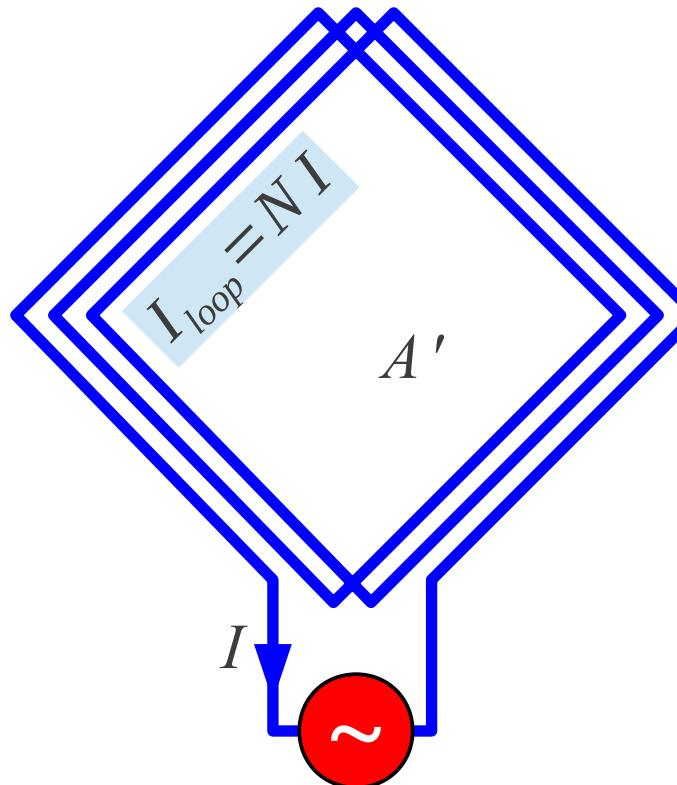
Small magnetic dipole



$$R = \frac{Z k^4 (A')^2}{6\pi} = \frac{8\pi^3 Z}{3} \left(\frac{A'}{\lambda^2} \right)^2$$

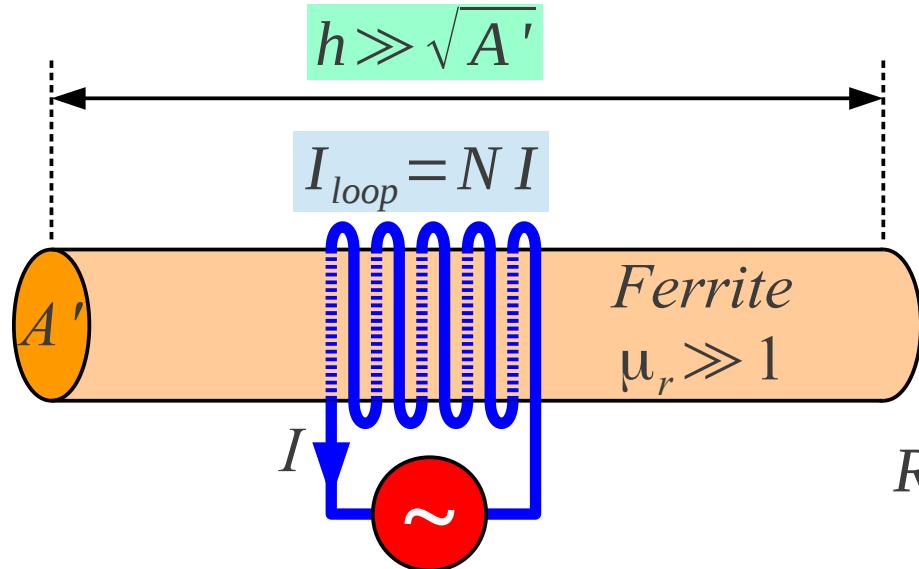
$$R = \frac{Z k^4 (N A')^2}{6\pi} = \frac{8\pi^3 Z}{3} \left(\frac{N A'}{\lambda^2} \right)^2$$

Loop antenna ~ 1920



$$\begin{aligned} f &\approx 300 \text{ kHz} \\ A' &\approx 1 \text{ m}^2 \\ N &\approx 10 \\ Z_0 &= \sqrt{\frac{\mu_0}{\epsilon_0}} \approx 377 \Omega \\ \lambda &= c_0/f = 1 \text{ km} \\ R_S &\approx 3.1 \mu\Omega \end{aligned}$$

Ferrite antenna ~ 1970



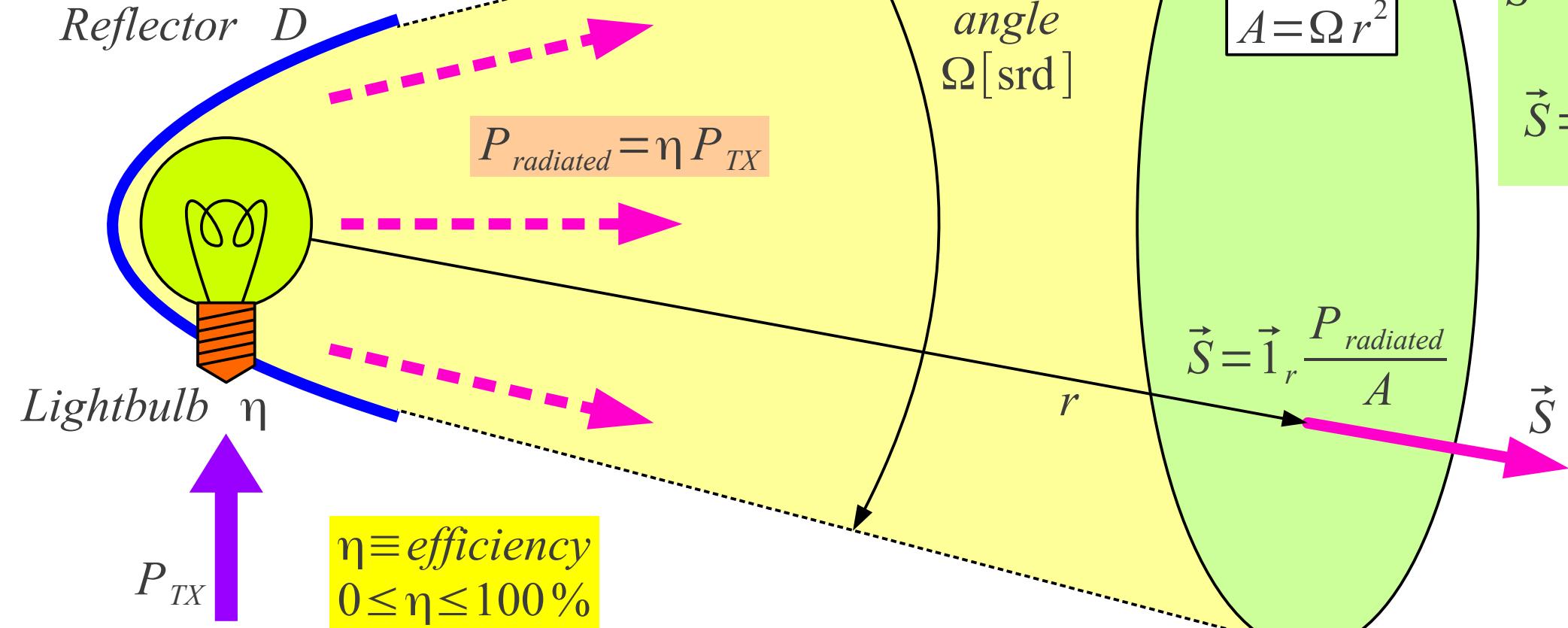
$$R = \frac{Z k^4 (\mu_r N A')^2}{6\pi} = \frac{8\pi^3 Z}{3} \left(\frac{\mu_r N A'}{\lambda^2} \right)^2$$

$$\begin{aligned} f &\approx 1 \text{ MHz} \\ A' &\approx 1 \text{ cm}^2 \\ h &\approx 20 \text{ cm} \\ \mu_r &\approx 100 \\ N &\approx 30 \\ Z_0 &= \sqrt{\frac{\mu_0}{\epsilon_0}} \approx 377 \Omega \\ \lambda &= c_0/f = 300 \text{ m} \\ R &\approx 0.35 \mu\Omega \end{aligned}$$

$$D = \frac{4\pi}{\Omega} \equiv \text{directivity}$$

$$D[\text{dBi}] = 10 \log_{10} D$$

Free space
 μ_0 ϵ_0
 loss-less!



$$\vec{S} = \vec{l}_r \frac{\eta P_{TX}}{\Omega r^2}$$

$$\vec{S} = \vec{l}_r \frac{\eta D P_{TX}}{4\pi r^2}$$

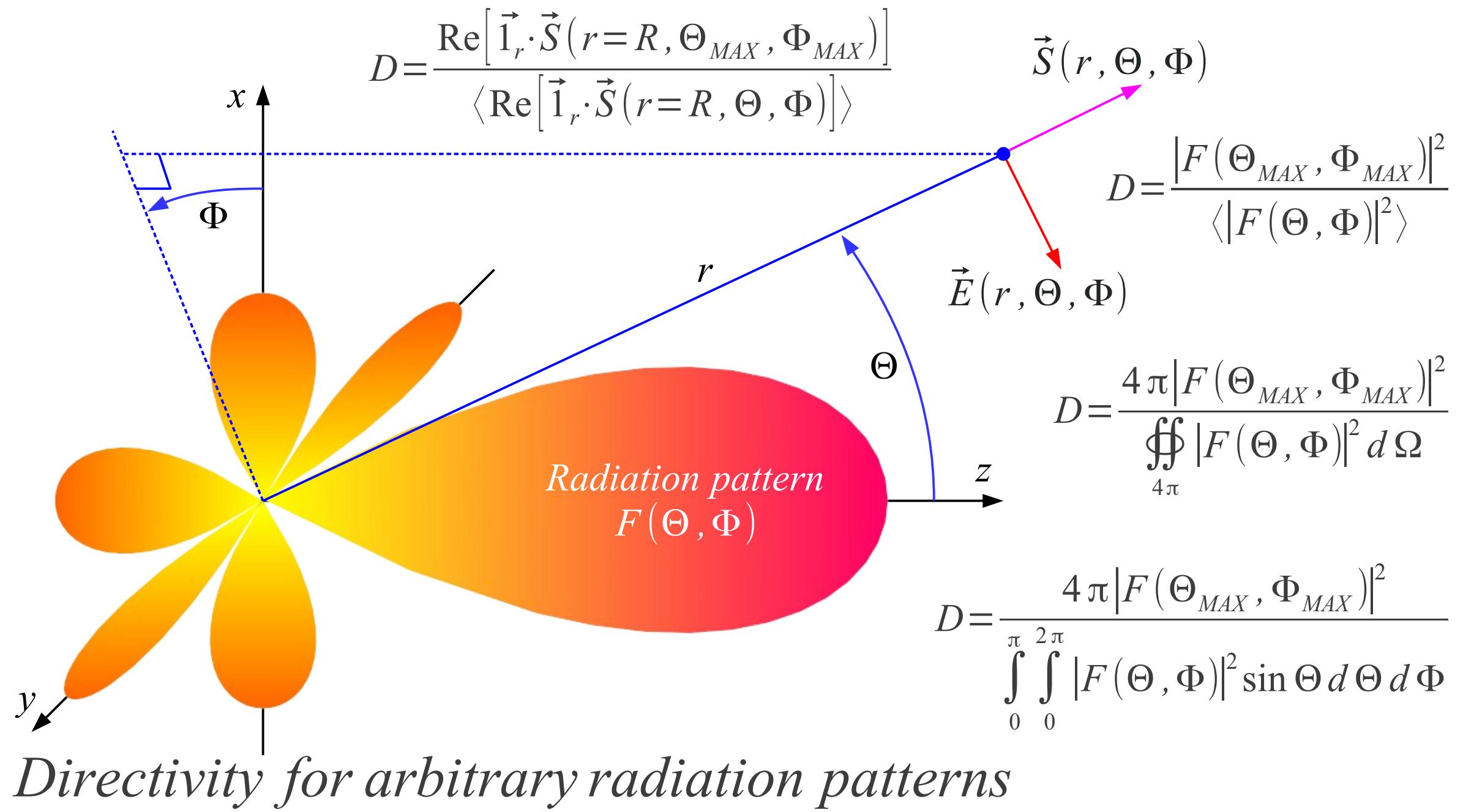
$$\vec{S} = \vec{l}_r \frac{G P_{TX}}{4\pi r^2}$$

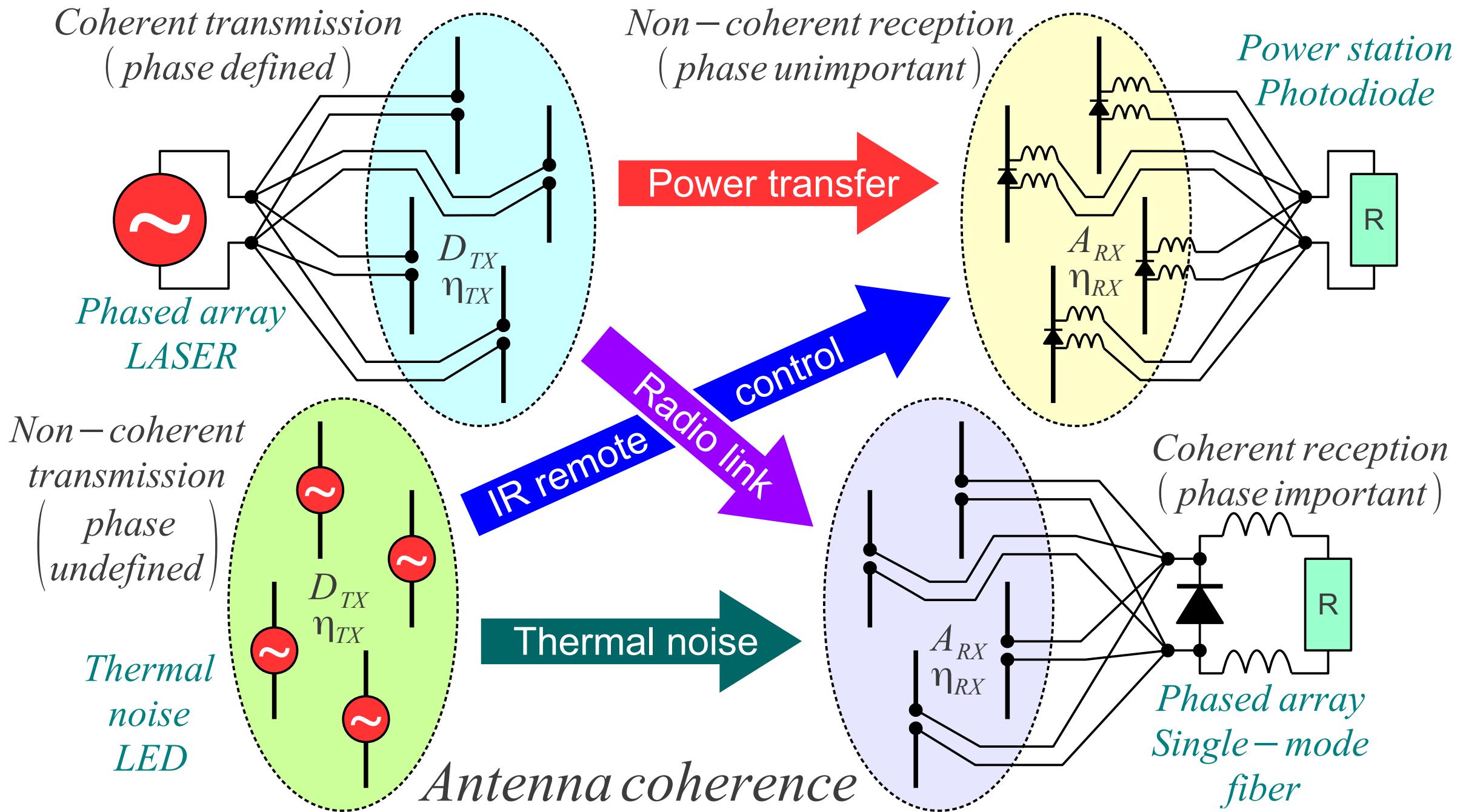
$$G = \eta D \equiv \text{gain}$$

$$G[\text{dBi}] = 10 \log_{10} G$$

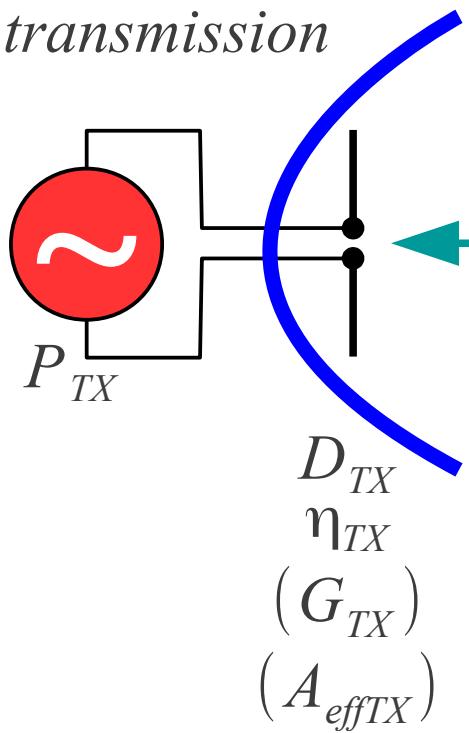
$$EIRP = D P_{radiated} = G P_{TX}$$

Directional transmitter





*Coherent
transmission*



*Free space
 $\mu_0 \quad \epsilon_0$
loss-less!*

r (*far field?*)

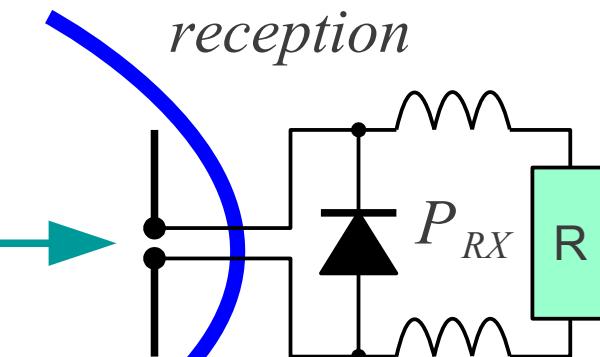
Harald Friis 1945

$$P_{RX} = P_{TX} \frac{\eta_{TX} D_{TX} A_{effRX} \eta_{RX}}{4 \pi r^2}$$

Antenna gains: $P_{RX} = P_{TX} G_{TX} G_{RX} \left(\frac{\lambda}{4 \pi r} \right)^2$

Antenna sizes: $P_{RX} = P_{TX} \frac{\eta_{TX} A_{effTX} A_{effRX} \eta_{RX}}{\lambda^2 r^2}$

*Coherent
reception*



A_{effRX}
 η_{RX}
 (D_{RX})
 (G_{RX})

Coherent antenna

$$D = \frac{4\pi}{\lambda^2} A_{eff}$$

$$G = \frac{4\pi}{\lambda^2} \eta A_{eff}$$

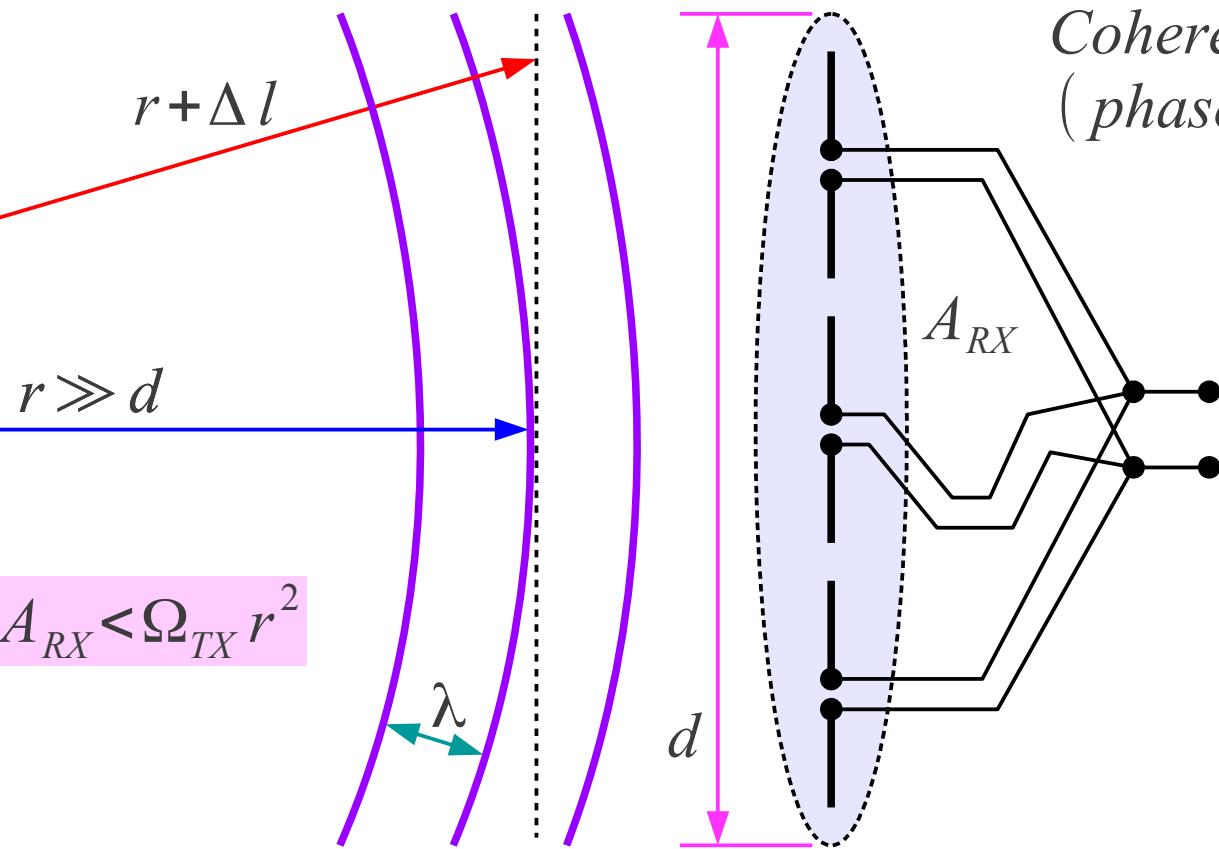
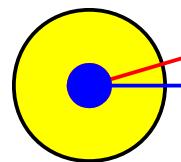
Reciprocity!

Friis equation

$$\Delta l = \sqrt{r^2 + (d/2)^2} - r \approx d^2/8r$$

$$\Delta\phi = k \Delta l$$

Point source



*Coherent reception
(phase important)*

Antenna far field
 $r \geq \frac{2d^2}{\lambda}$
 $F(\Theta, \Phi)$
 $D \quad G$

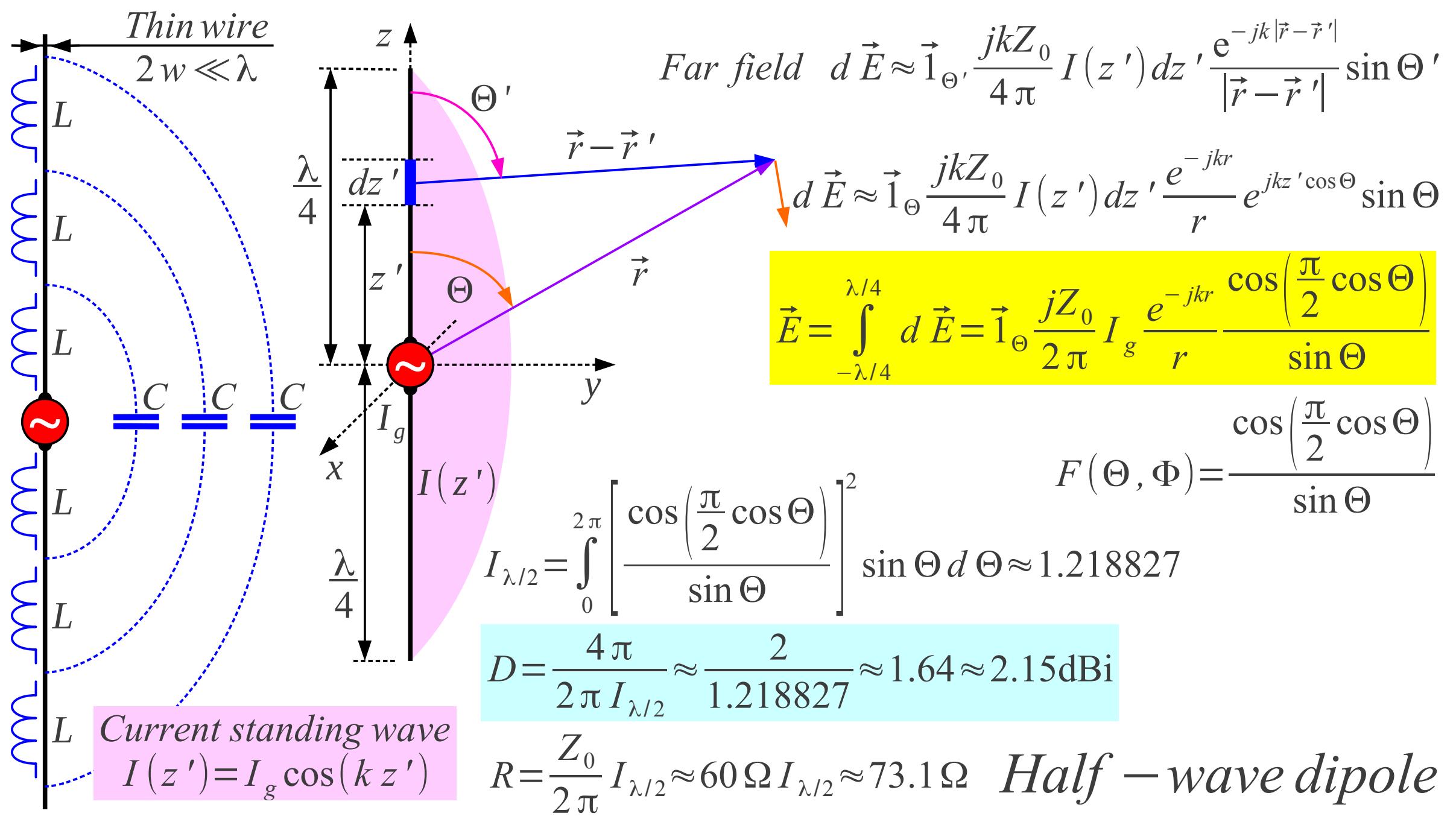
Phase tighter than magnitude $A_{RX} < \Omega_{TX} r^2$

$$\Delta P_{dB} \approx 20 \log_{10} \left| \frac{\sin(\Delta\phi/2)}{\Delta\phi/2} \right|$$

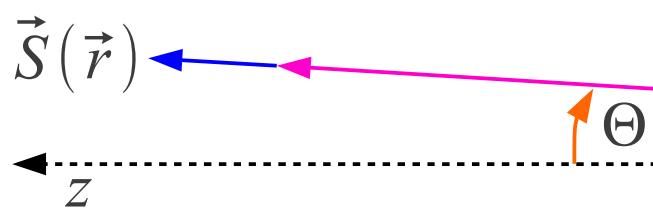
Δl	$\Delta\phi$ [rd]	ΔP [dB]	$r \geq$	Usage
$\lambda/2$	π	-3.922	$d^2/4\lambda$	Photo depth of field
$\lambda/4$	$\pi/2$	-0.912	$d^2/2\lambda$	Lord Rayleigh 1891
$\lambda/8$	$\pi/4$	-0.224	d^2/λ	
$\lambda/16$	$\pi/8$	-0.056	$2d^2/\lambda$	Antenna measurements

Example photo camera
Lens $\equiv d = 2$ mm
 $\lambda = 0.5 \mu m$
 $DoF \equiv d^2/4\lambda = 2$ m

Rayleigh distance



$$\vec{S} = \vec{l}_r \frac{|E|^2}{2Z_0} = \vec{l}_r \frac{(1+\cos\Theta)^2}{8Z_0\lambda^2 r^2} \left| \iint_A E_0(x, y) e^{jkx \sin\Theta \cos\Phi} e^{jky \sin\Theta \sin\Phi} dA \right|^2$$



$$\Theta_{MAX}=0 \rightarrow \cos\Theta=1 \quad \sin\Theta=0$$

$$\vec{S}_{MAX} = \vec{l}_r \frac{1}{2Z_0\lambda^2 r^2} \left| \iint_A E_0(x, y) dA \right|^2$$

$$D = \frac{|\vec{S}_{MAX}|}{P/(4\pi r^2)} = \frac{4\pi}{\lambda^2} \frac{\left| \iint_A E_0(x, y) dA \right|^2}{\iint_A |E_0(x, y)|^2 dA}$$

$$Example \quad E_0(x, y) = const. \rightarrow D = \frac{4\pi}{\lambda^2} A$$

Aperture radiation

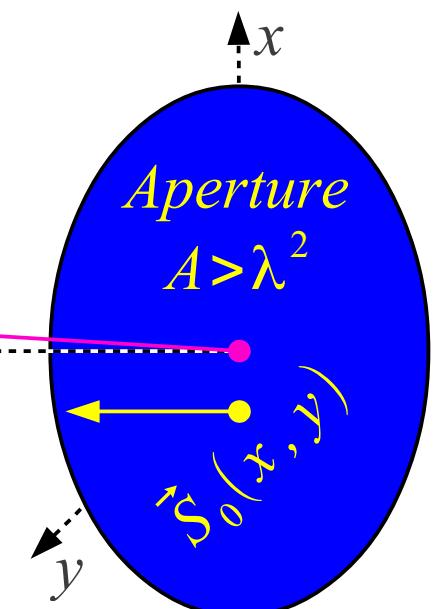
$$P = \iint_A \vec{S}_0 \cdot \vec{l}_z dA = \iint_A \frac{|E_0(x, y)|^2}{2Z_0} dA$$

$A_{eff} = \eta_0 A \equiv effective\ area$

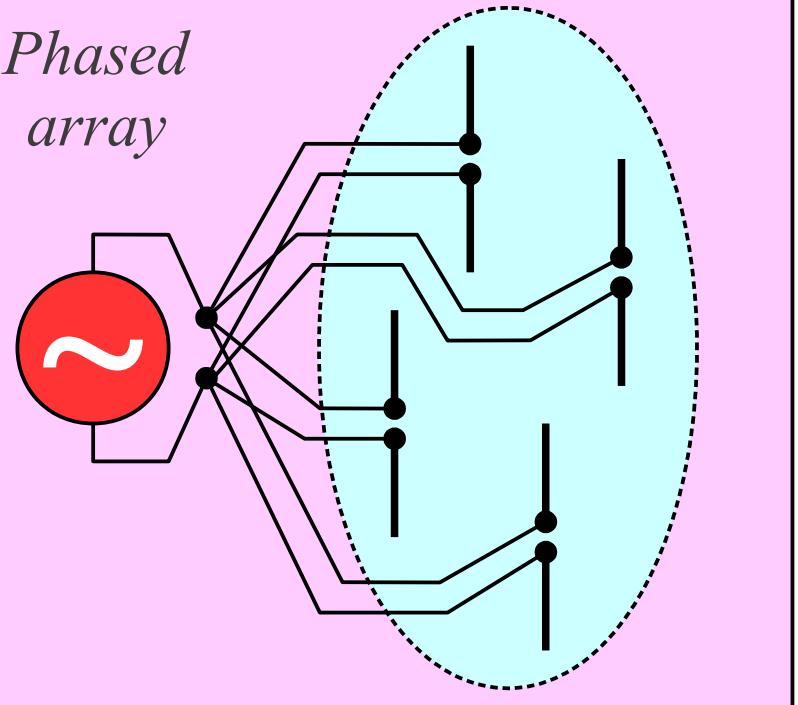
$$D = \frac{4\pi}{\lambda^2} A_{eff} = \frac{4\pi}{\lambda^2} \eta_0 A$$

$$A_{eff} = \frac{\left| \iint_A E_0(x, y) dA \right|^2}{\iint_A |E_0(x, y)|^2 dA}$$

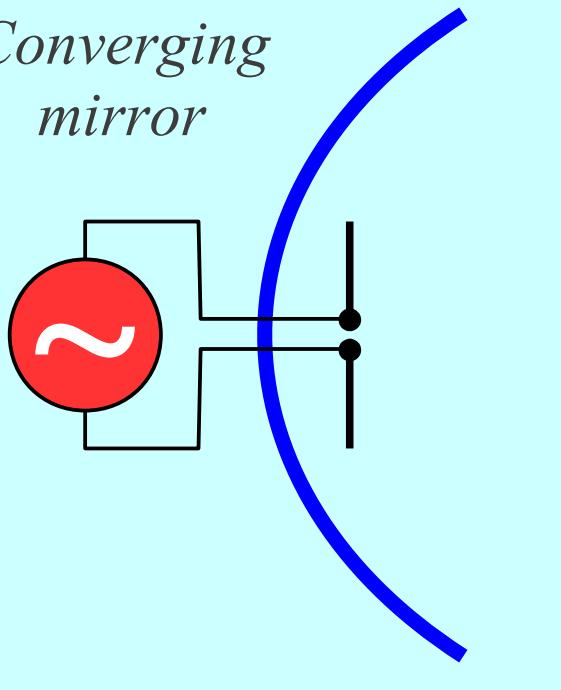
$$\eta_0 = \frac{\left| \iint_A E_0(x, y) dA \right|^2}{A \iint_A |E_0(x, y)|^2 dA} \equiv \begin{matrix} illumination \\ efficiency \end{matrix}$$



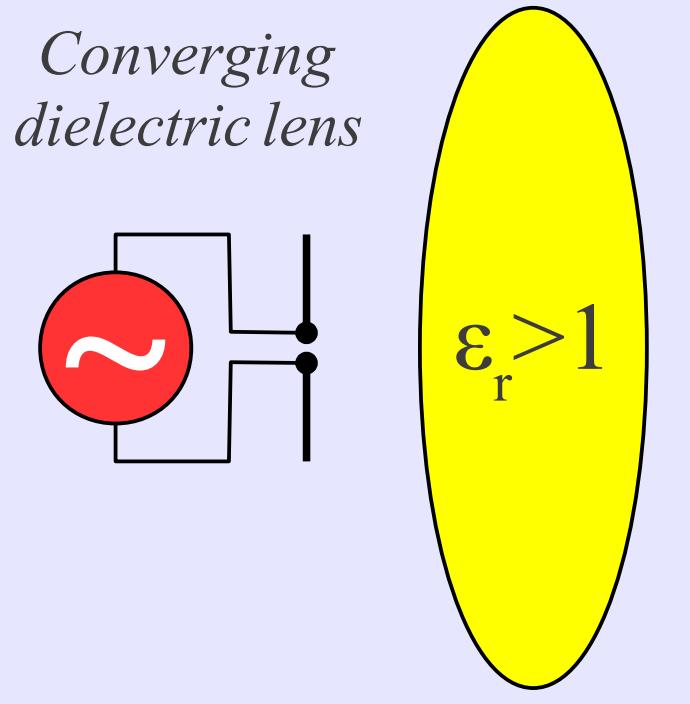
Phased array



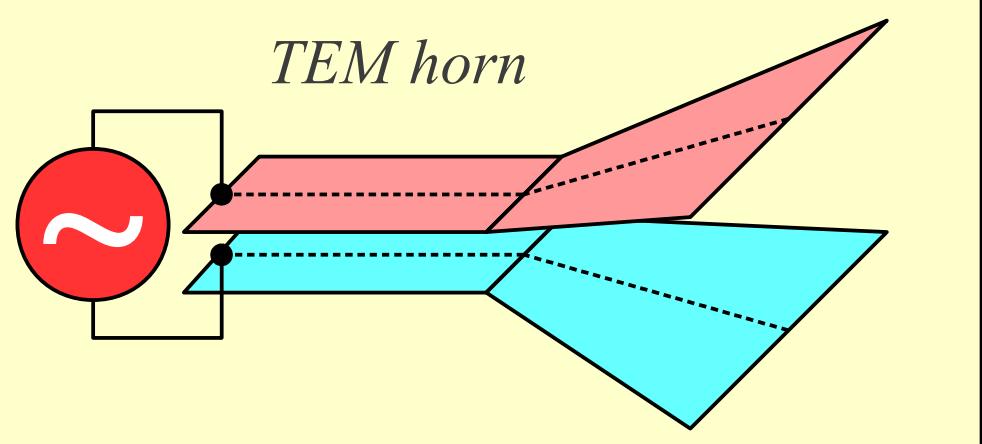
Converging mirror



Converging dielectric lens

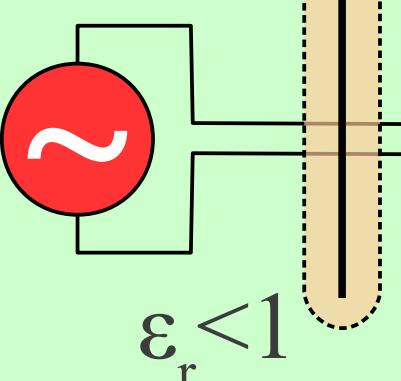


TEM horn



Aperture directional antennas

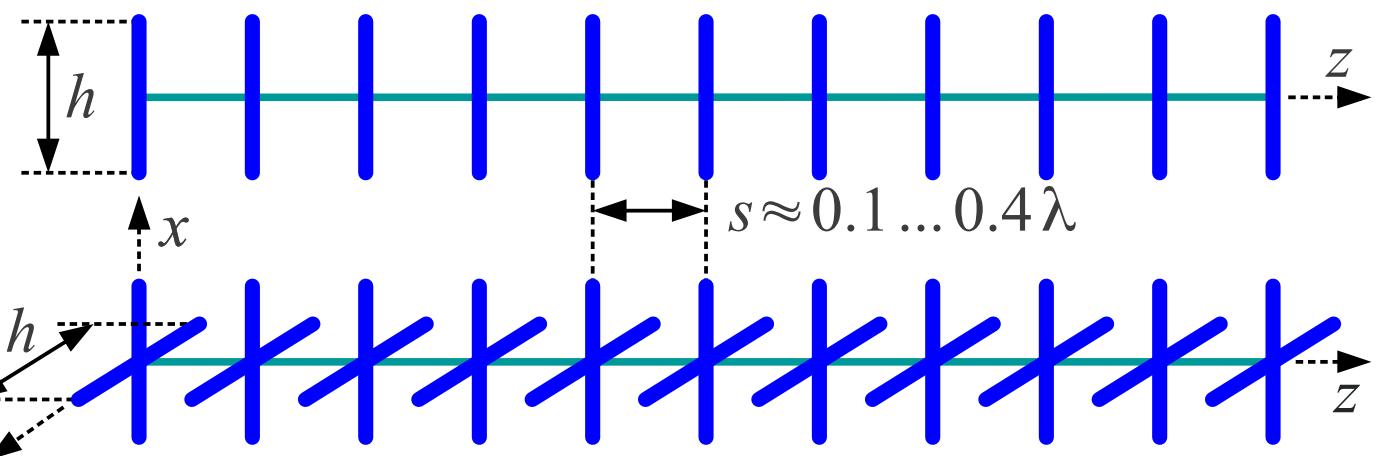
Diverging lens



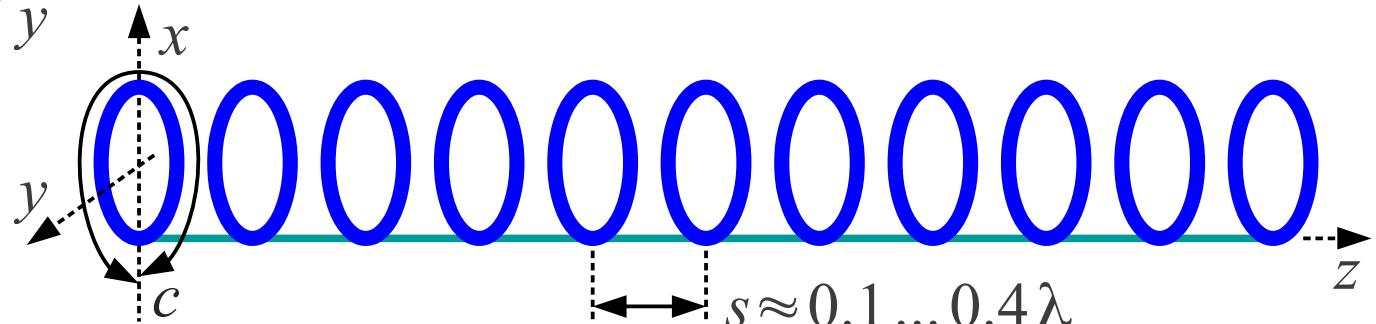
*Artificial dielectric
converging lens*



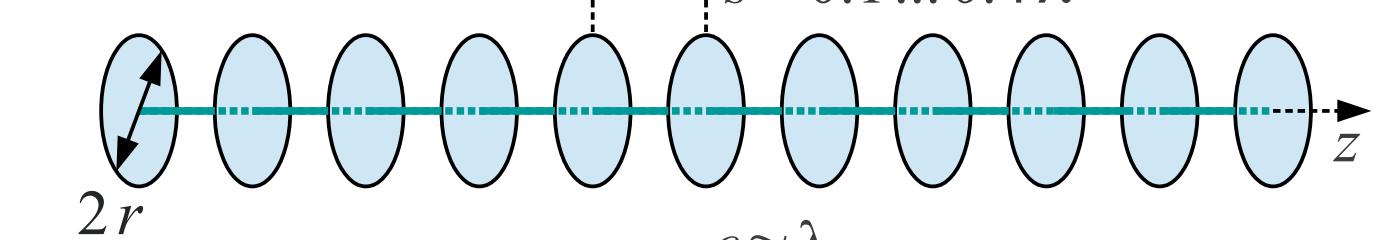
Metal rods $h \approx 0.45\lambda$
(Shintaro Uda 1926)



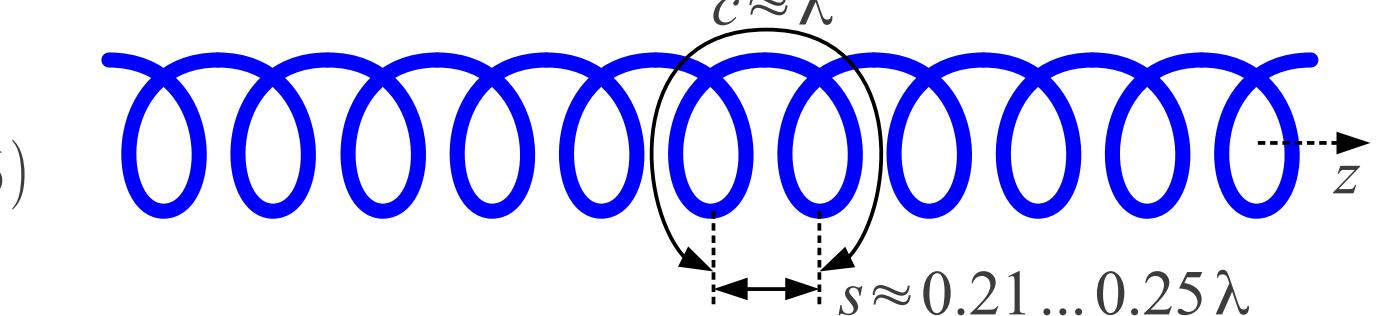
Crossed metal rods
(both polarizations)



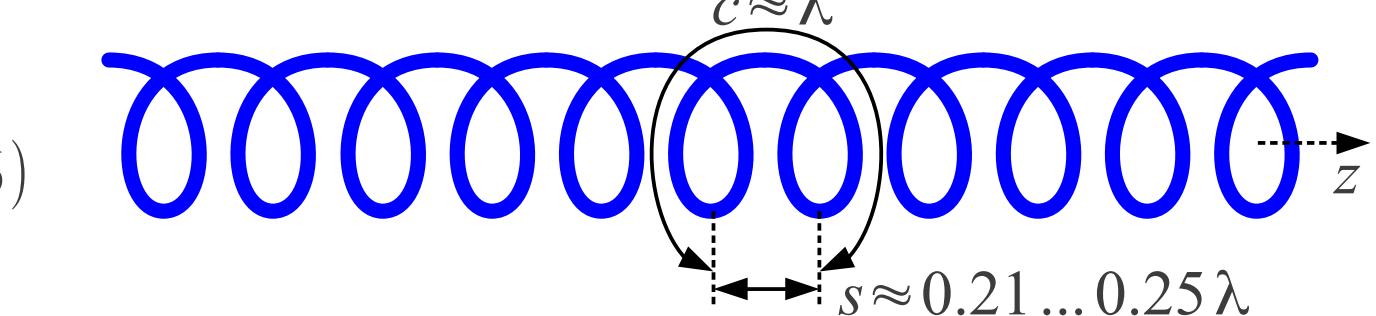
Wire loops $c \approx 0.9\lambda$
(different shapes, loop-Yagi)



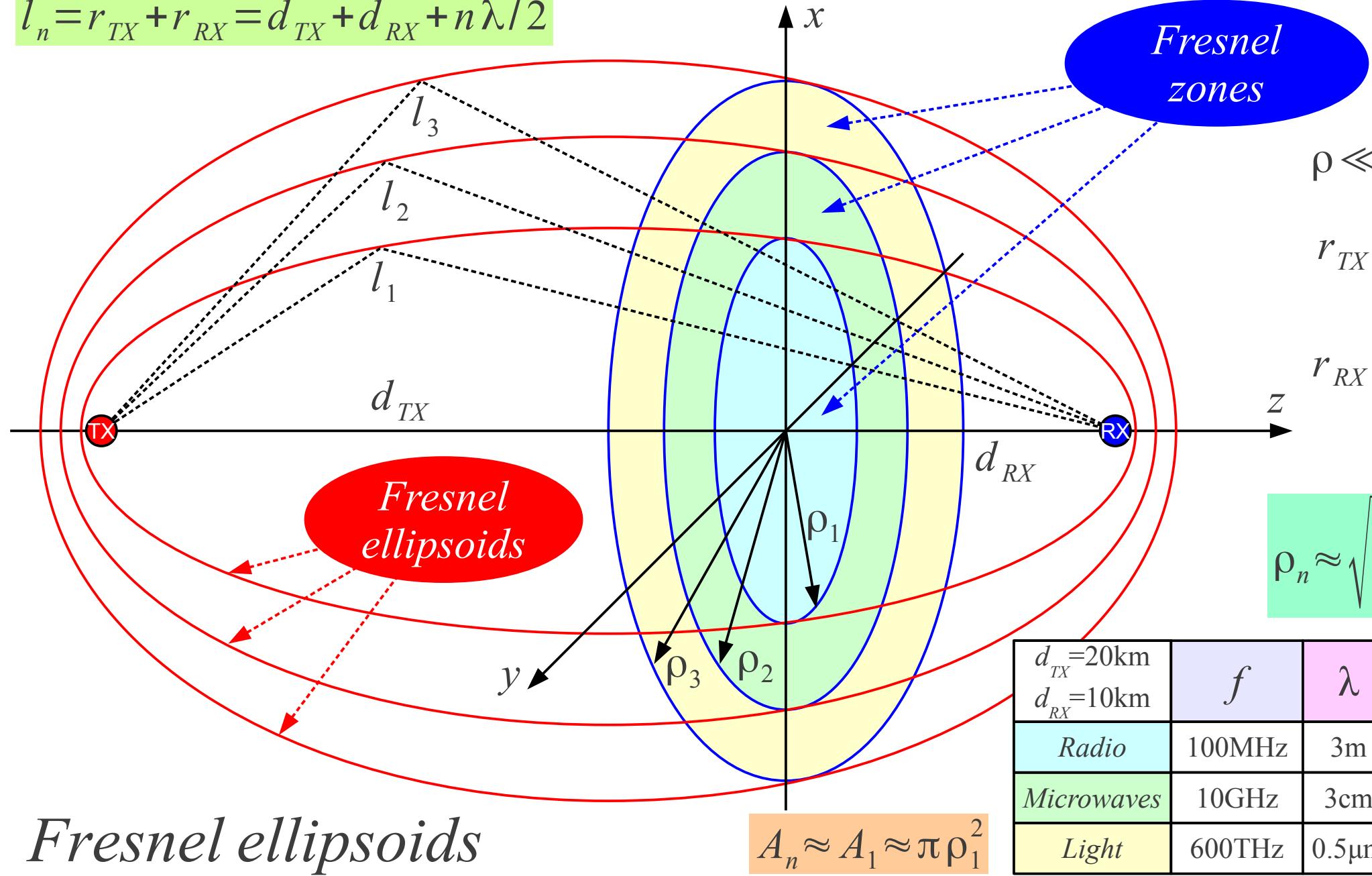
Metal disks $2r \approx 0.3\lambda$
(both polarizations, disk-Yagi)
(J.C.Simon & G.Weil 1953)



Helix $0.75\lambda < c < 1.33\lambda$
(circular polarization, J.Kraus 1946)
Slow-wave structures



$$l_n = r_{TX} + r_{RX} = d_{TX} + d_{RX} + n\lambda/2$$



$$\rho \ll d_{TX}, d_{RX}$$

$$r_{TX} \approx d_{TX} + \frac{\rho^2}{2d_{TX}}$$

$$r_{RX} \approx d_{RX} + \frac{\rho^2}{2d_{RX}}$$

$$\rho_n \approx \sqrt{n\lambda \frac{d_{TX}d_{RX}}{d_{TX}+d_{RX}}}$$

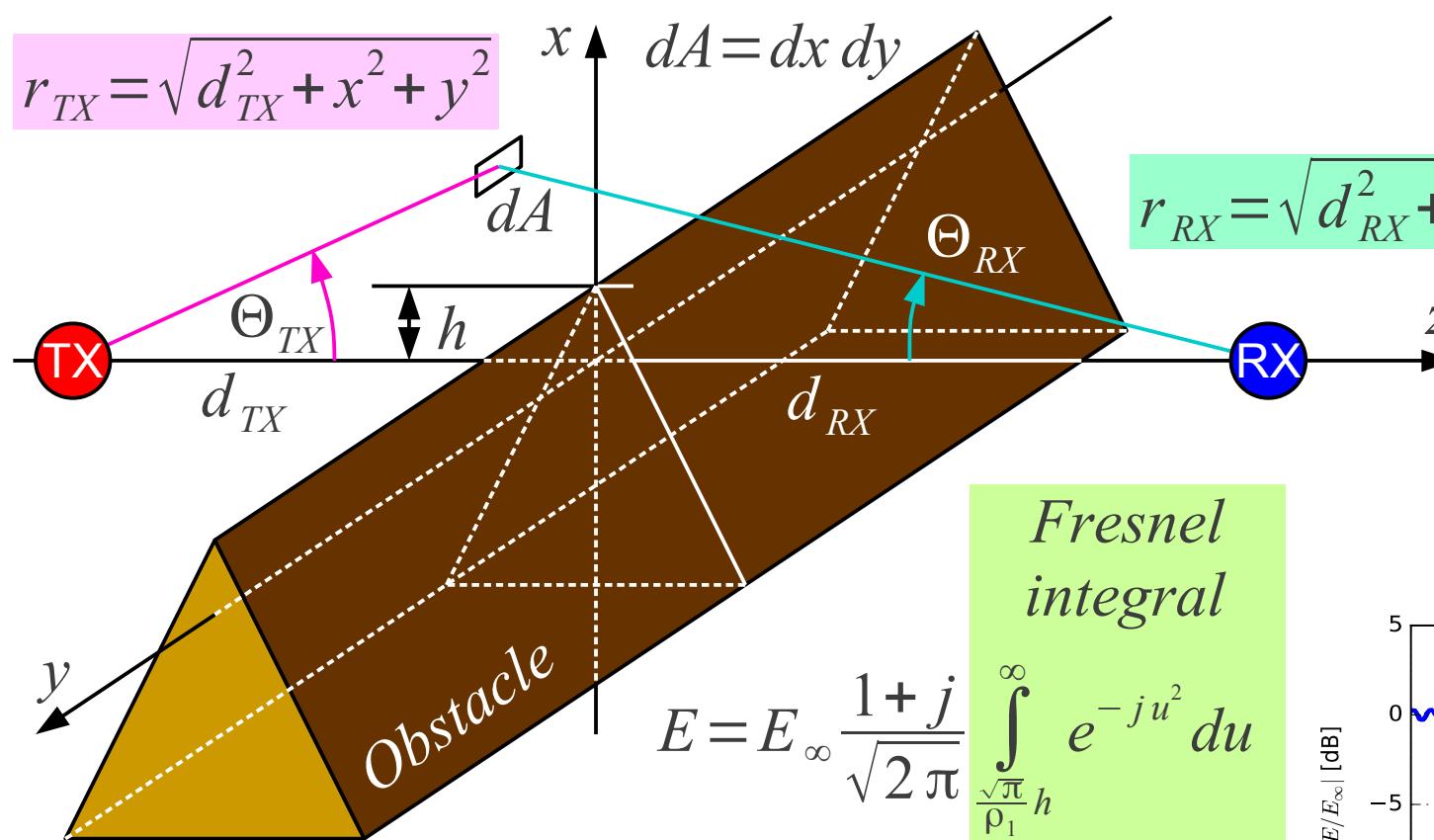
	d_{TX} km	f	λ	ρ_1	A_1
Radio	20	100MHz	3m	141m	62831m ²
Microwaves	10	10GHz	3cm	14.1m	628m ²
Light		600THz	0.5μm	5.8cm	0.0105m ²

Fresnel ellipsoids

$$A_n \approx A_1 \approx \pi \rho_1^2$$

Knife-edge diffraction

$$r_{TX} = \sqrt{d_{TX}^2 + x^2 + y^2}$$



$$\begin{aligned} x, y &\ll d_{TX}, d_{RX} \\ \cos \Theta_{TX} &\approx 1 \\ \cos \Theta_{RX} &\approx 1 \end{aligned}$$

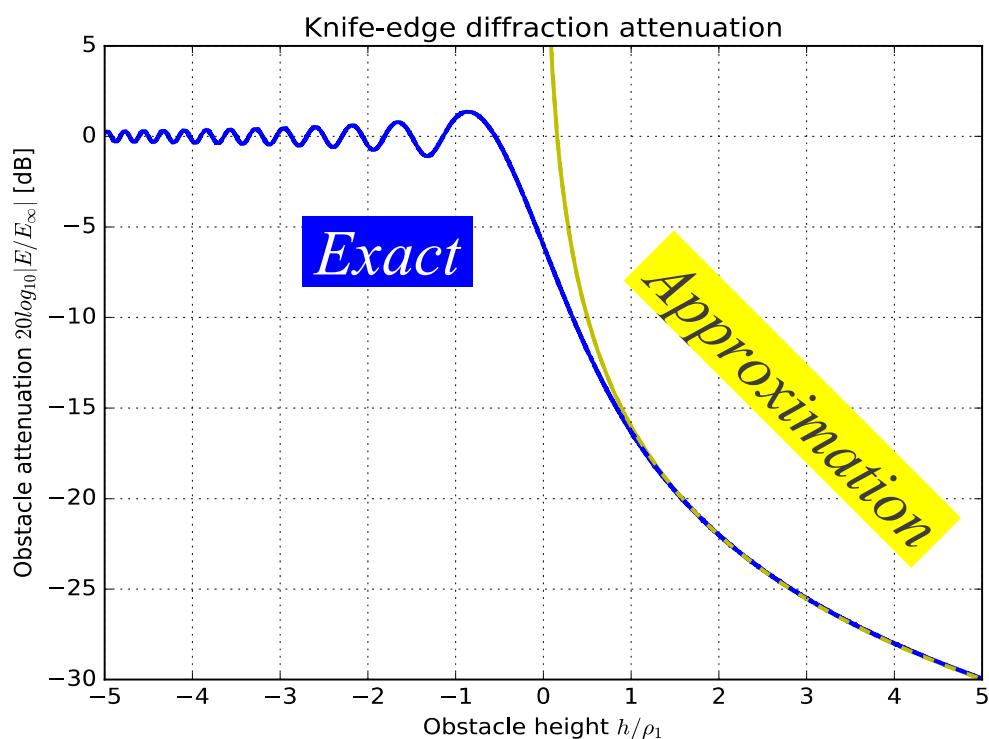
$$a_{dB} = 20 \log_{10} \frac{1}{\sqrt{\pi}} \left| \int_{\frac{\sqrt{\pi}}{\rho_1} h}^{\infty} e^{-ju^2} du \right|$$

Approximation $h \geq \rho_1$

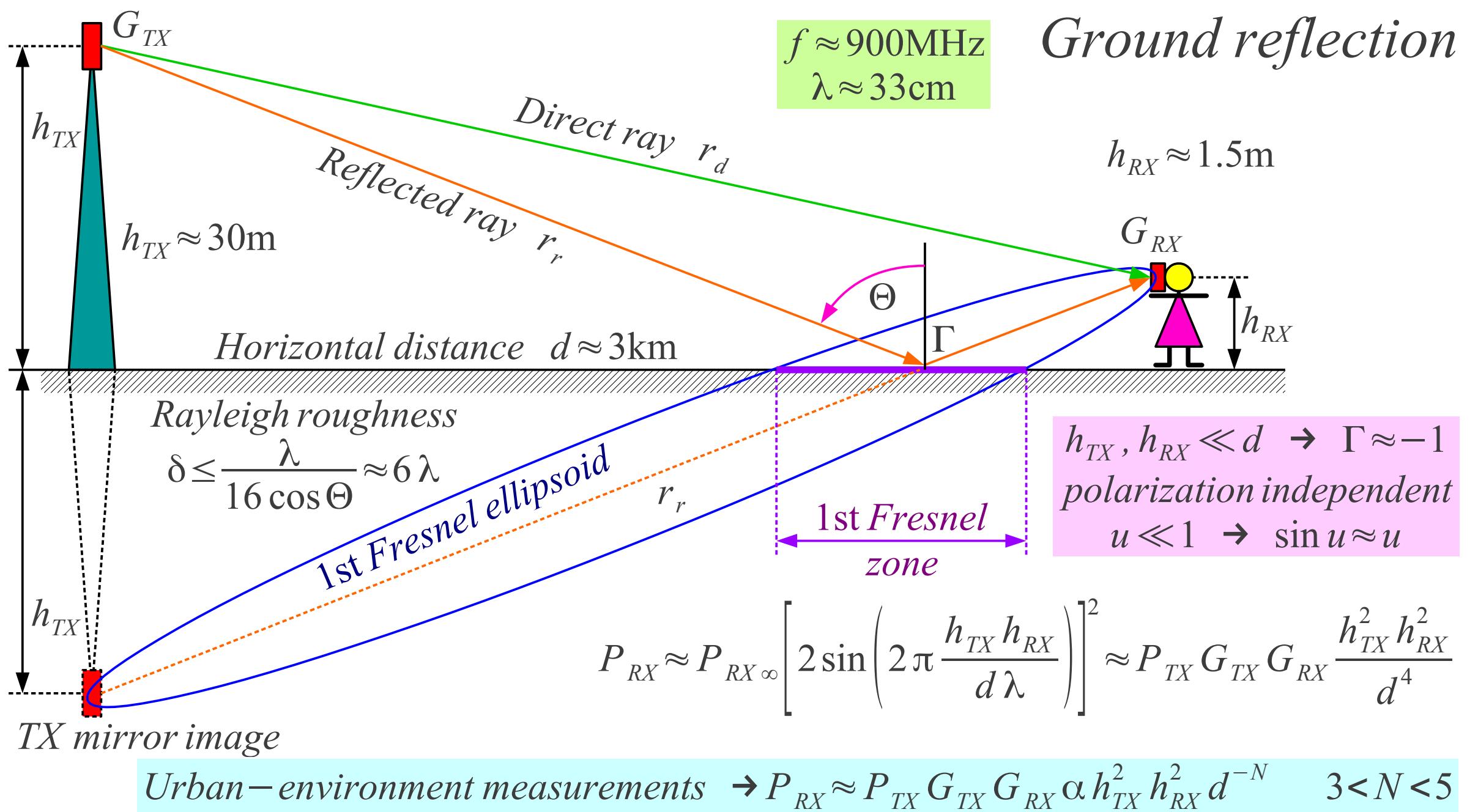
$$a_{dB} \approx -16 - 20 \log_{10} \frac{h}{\rho_1}$$

$$\rho_1 \approx \sqrt{\lambda} \frac{d_{TX} d_{RX}}{d_{TX} + d_{RX}}$$

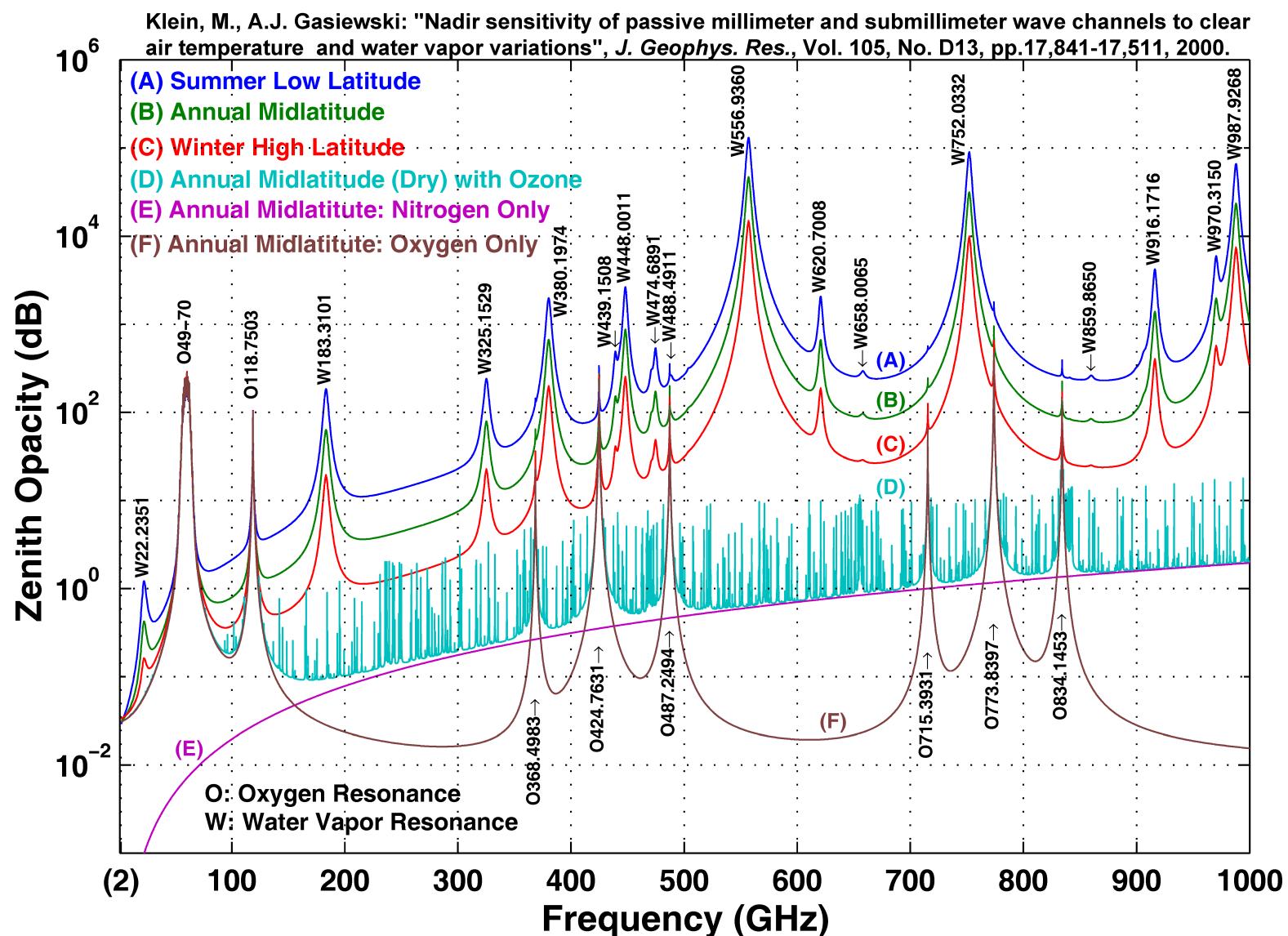
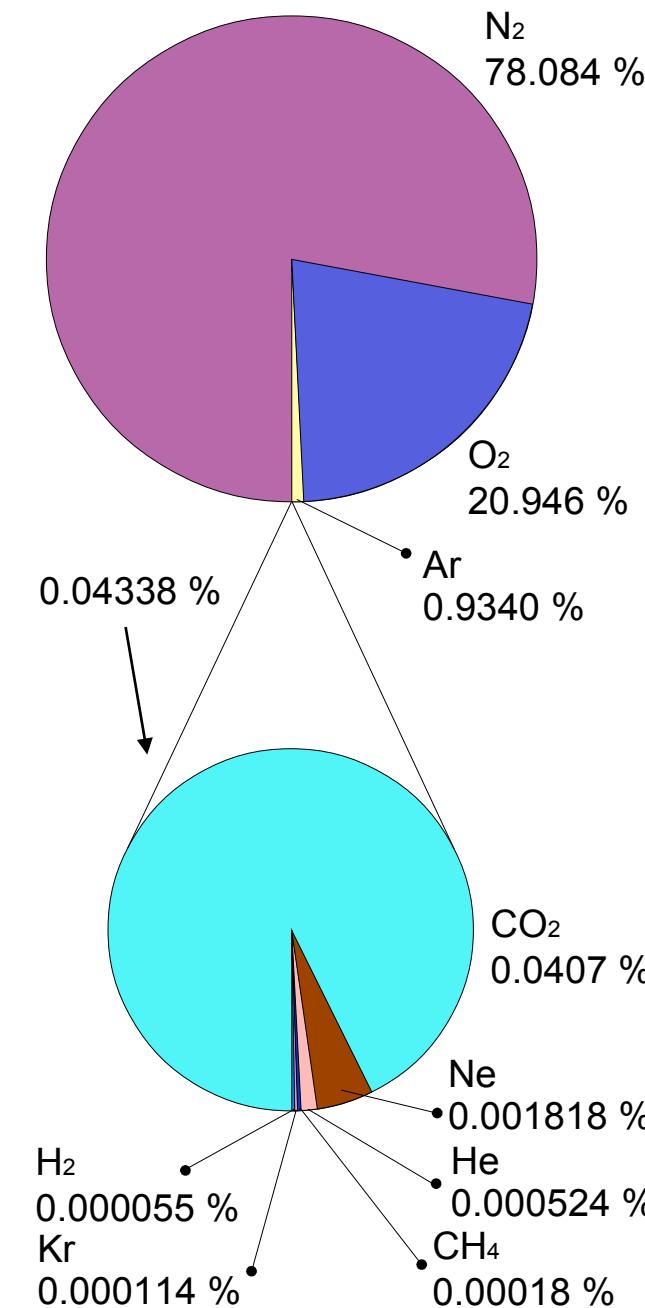
$$E = E_\infty \frac{j}{\rho_1^2 h} \int_{-h}^{\infty} e^{-j\pi \frac{x^2}{\rho_1^2}} dx \int_{-\infty}^{\infty} e^{-j\pi \frac{y^2}{\rho_1^2}} dy$$



Ground reflection

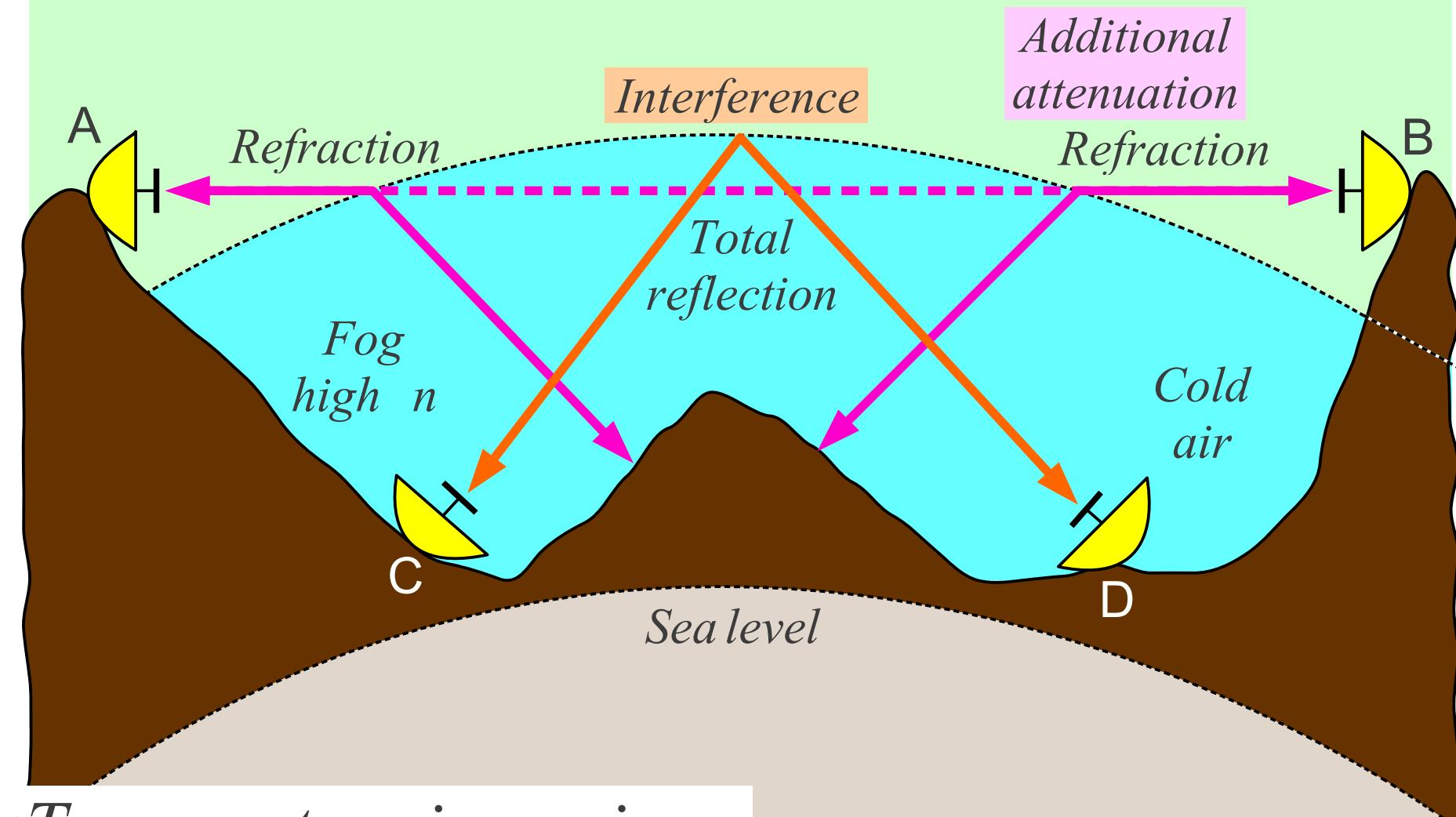


Atmospheric attenuation



Drawing
not to scale

Warm air \rightarrow low n



Temperature inversion

